

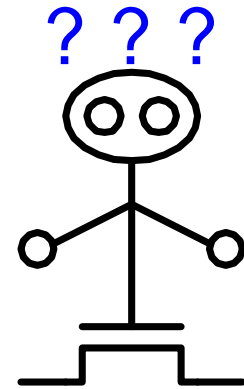
# Lecture 6: Logical Effort

# Outline

- Logical Effort
- Delay in a Logic Gate
- Multistage Logic Networks
- Choosing the Best Number of Stages
- Example
- Summary

# Introduction

- ❑ Chip designers face a bewildering array of choices
  - What is the best circuit topology for a function?
  - How many stages of logic give least delay?
  - How wide should the transistors be?
- ❑ Logical effort is a method to make these decisions
  - Uses a simple model of delay
  - Allows back-of-the-envelope calculations
  - Helps make rapid comparisons between alternatives
  - Emphasizes remarkable symmetries



# Example

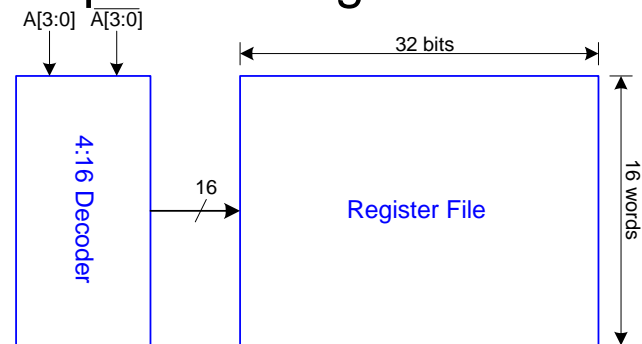
- ❑ Ben Bitdiddle is the memory designer for the Motorola 68W86, an embedded automotive processor. Help Ben design the decoder for a register file.

- ❑ Decoder specifications:

- 16 word register file
- Each word is 32 bits wide
- Each bit presents load of 3 unit-sized transistors
- True and complementary address inputs  $A[3:0]$
- Each input may drive 10 unit-sized transistors

- ❑ Ben needs to decide:

- How many stages to use?
- How large should each gate be?
- How fast can decoder operate?

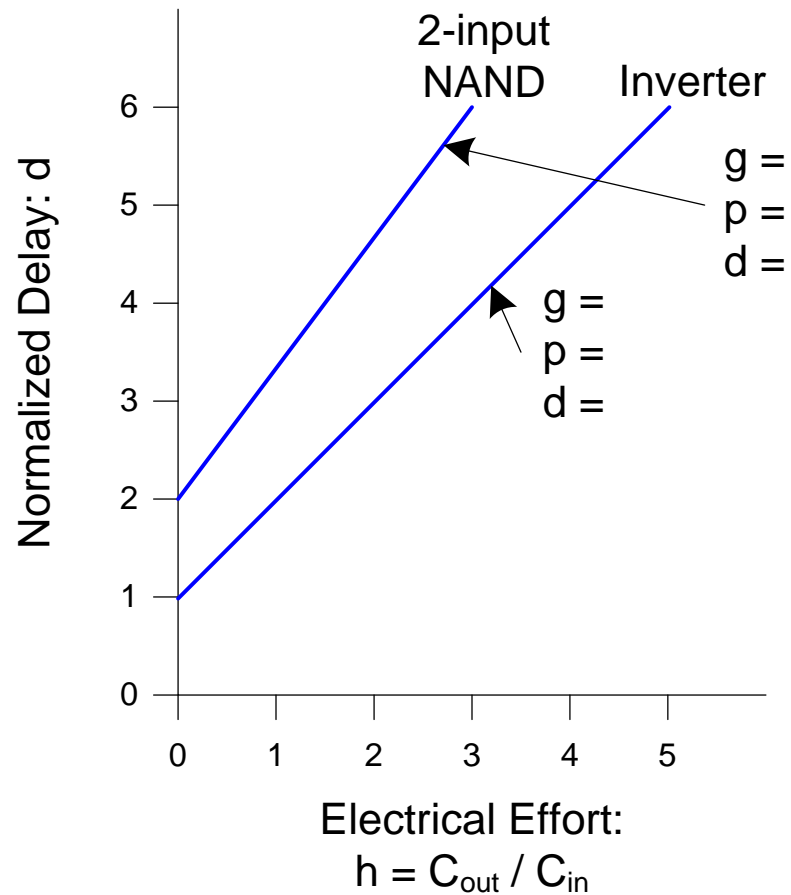


# Delay in a Logic Gate

- ❑ Express delays in process-independent unit  $d = \frac{d_{abs}}{\tau}$
  - ❑ Delay has two components:  $d = f + p$
  - ❑  $f$ : *effort delay* =  $gh$  (a.k.a. stage effort)
    - Again has two components
  - ❑  $g$ : *logical effort*
    - Measures relative ability of gate to deliver current
    - $g \equiv 1$  for inverter
  - ❑  $h$ : *electrical effort* =  $C_{out} / C_{in}$ 
    - Ratio of output to input capacitance
    - Sometimes called fanout
  - ❑  $p$ : *parasitic delay*
    - Represents delay of gate driving no load
    - Set by internal parasitic capacitance
- $\tau = 3RC$   
 $\approx 3 \text{ ps in } 65 \text{ nm process}$   
 $60 \text{ ps in } 0.6 \mu\text{m process}$

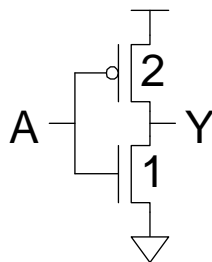
# Delay Plots

$$d = f + p$$
$$= gh + p$$

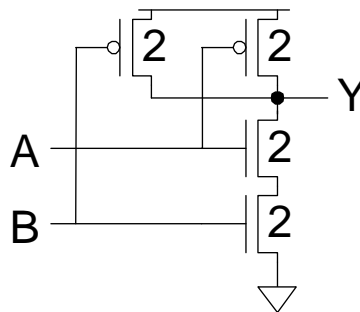


# Computing Logical Effort

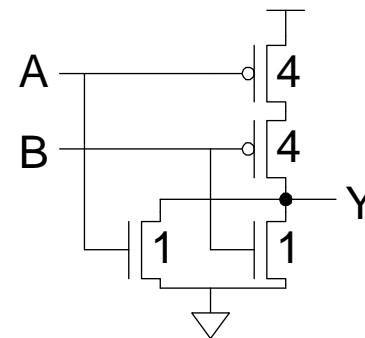
- ❑ DEF: *Logical effort is the ratio of the input capacitance of a gate to the input capacitance of an inverter delivering the same output current.*
- ❑ Measure from delay vs. fanout plots
- ❑ Or estimate by counting transistor widths



$$C_{in} = 3$$
$$g = 3/3$$



$$C_{in} = 4$$
$$g = 4/3$$



$$C_{in} = 5$$
$$g = 5/3$$

# Catalog of Gates

## □ Logical effort of common gates

Gate type	Number of inputs				
	1	2	3	4	n
Inverter	1				
NAND		$4/3$	$5/3$	$6/3$	$(n+2)/3$
NOR		$5/3$	$7/3$	$9/3$	$(2n+1)/3$
Tristate / mux	2	2	2	2	2
XOR, XNOR		4, 4	6, 12, 6	8, 16, 16, 8	



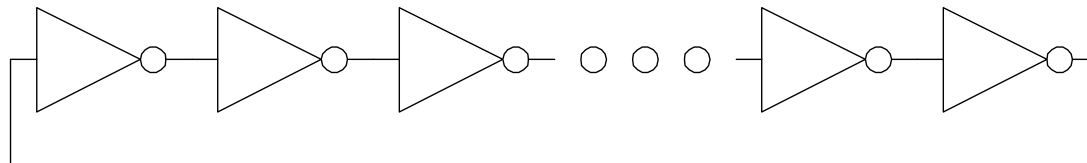
# Catalog of Gates

- Parasitic delay of common gates
  - In multiples of  $p_{inv}$  ( $\approx 1$ )

Gate type	Number of inputs				
	1	2	3	4	n
Inverter	1				
NAND		2	3	4	n
NOR		2	3	4	n
Tristate / mux	2	4	6	8	2n
XOR, XNOR		4	6	8	

# Example: Ring Oscillator

- Estimate the frequency of an N-stage ring oscillator



Logical Effort:  $g =$

Electrical Effort:  $h =$

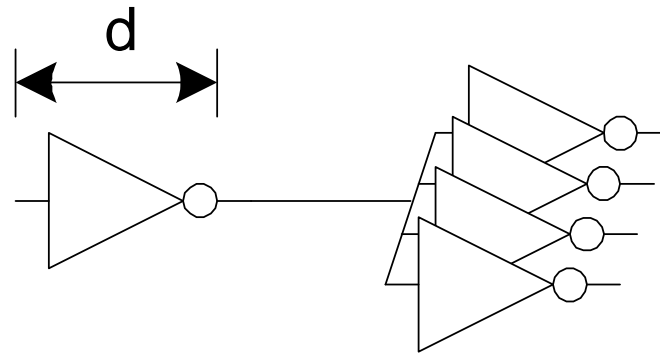
Parasitic Delay:  $p =$

Stage Delay:  $d =$

Frequency:  $f_{osc} =$

# Example: FO4 Inverter

- Estimate the delay of a fanout-of-4 (FO4) inverter



Logical Effort:  $g =$

Electrical Effort:  $h =$

Parasitic Delay:  $p =$

Stage Delay:  $d =$

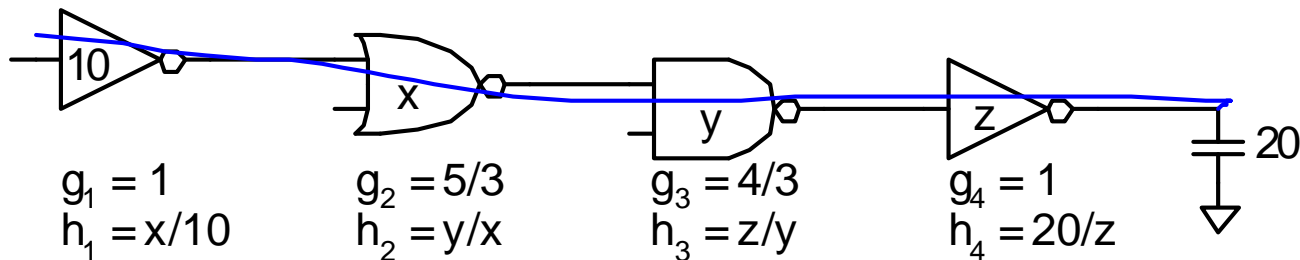
# Multistage Logic Networks

- ❑ Logical effort generalizes to multistage networks

- ❑ *Path Logical Effort*  $G = \prod g_i$

- ❑ *Path Electrical Effort*  $H = \frac{C_{\text{out-path}}}{C_{\text{in-path}}}$

- ❑ *Path Effort*  $F = \prod f_i = \prod g_i h_i$



# Multistage Logic Networks

❑ Logical effort generalizes to multistage networks

❑ *Path Logical Effort*  $G = \prod g_i$

❑ *Path Electrical Effort*  $H = \frac{C_{out-path}}{C_{in-path}}$

❑ *Path Effort*  $F = \prod f_i = \prod g_i h_i$

❑ Can we write  $F = GH$ ?

# Paths that Branch

❑ No! Consider paths that branch:

G =

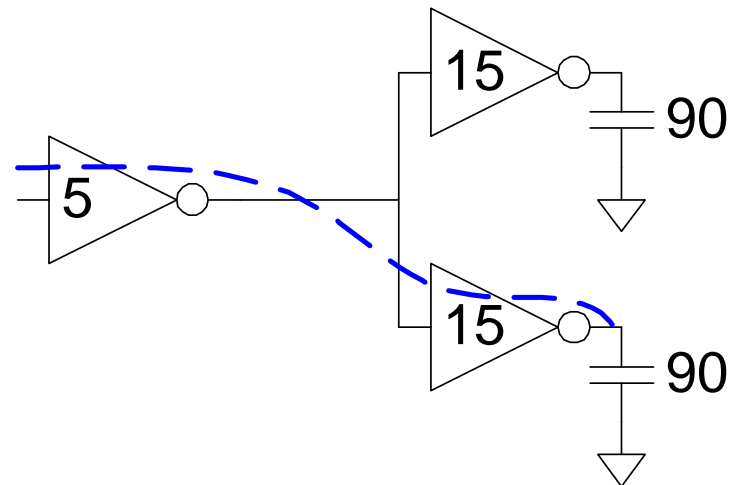
H =

GH =

$h_1$  =

$h_2$  =

F =



# Branching Effort

- Introduce *branching effort*
  - Accounts for branching between stages in path

$$b = \frac{C_{\text{on path}} + C_{\text{off path}}}{C_{\text{on path}}}$$

$$B = \prod b_i$$

Note:

$$\prod h_i = BH$$

- Now we compute the path effort
  - $F = GBH$

# Multistage Delays

- ❑ Path Effort Delay  $D_F = \sum f_i$
- ❑ Path Parasitic Delay  $P = \sum p_i$
- ❑ Path Delay  $D = \sum d_i = D_F + P$



# Designing Fast Circuits

$$D = \sum d_i = D_F + P$$

- Delay is smallest when each stage bears same effort

$$\hat{f} = g_i h_i = F^{\frac{1}{N}}$$

- Thus minimum delay of N stage path is



- This is a **key** result of logical effort
  - Find fastest possible delay
  - Doesn't require calculating gate sizes

# Gate Sizes

- How wide should the gates be for least delay?

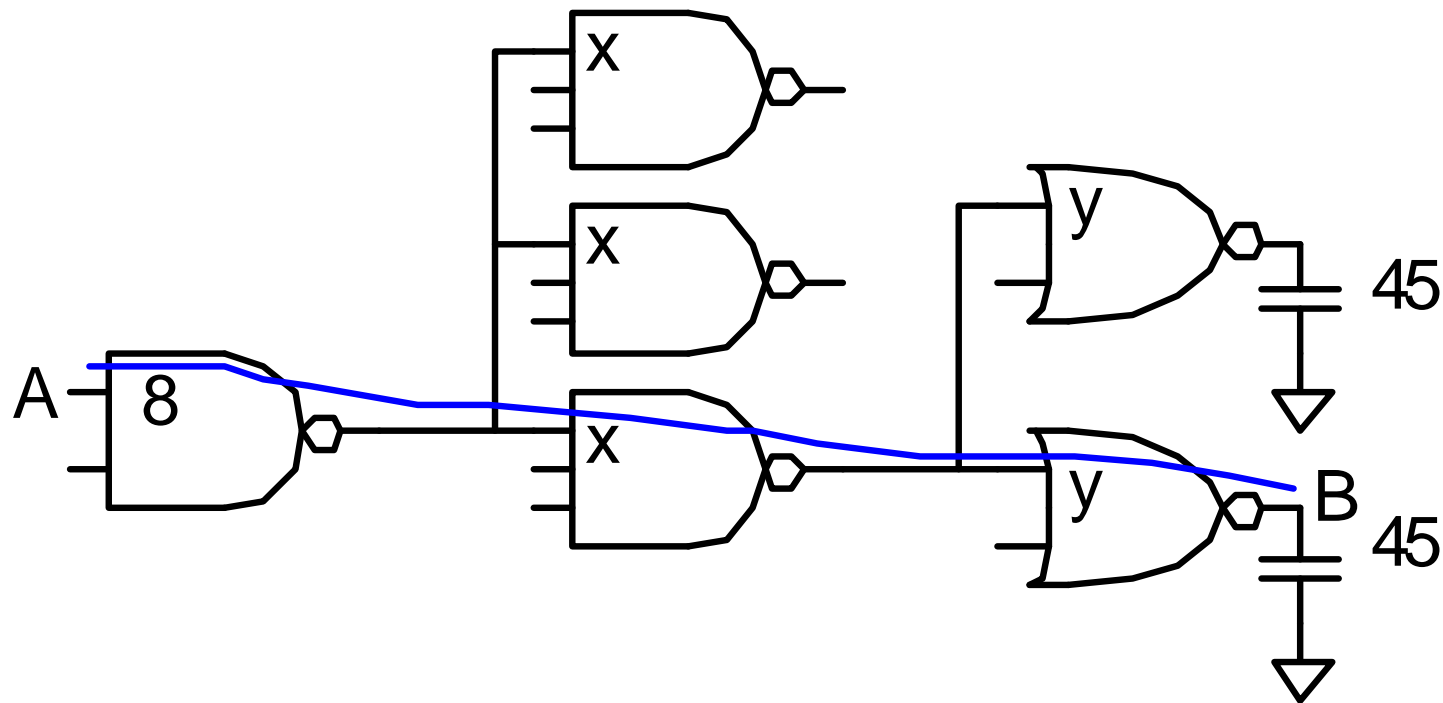
$$\hat{f} = gh = g \frac{C_{out}}{C_{in}}$$

$$\Rightarrow C_{in_i} = \frac{g_i C_{out_i}}{\hat{f}}$$

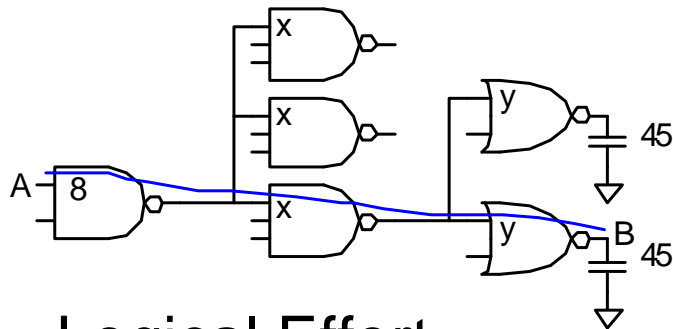
- Working backward, apply capacitance transformation to find input capacitance of each gate given load it drives.
- Check work by verifying input cap spec is met.

# Example: 3-stage path

- Select gate sizes  $x$  and  $y$  for least delay from  $A$  to  $B$



# Example: 3-stage path



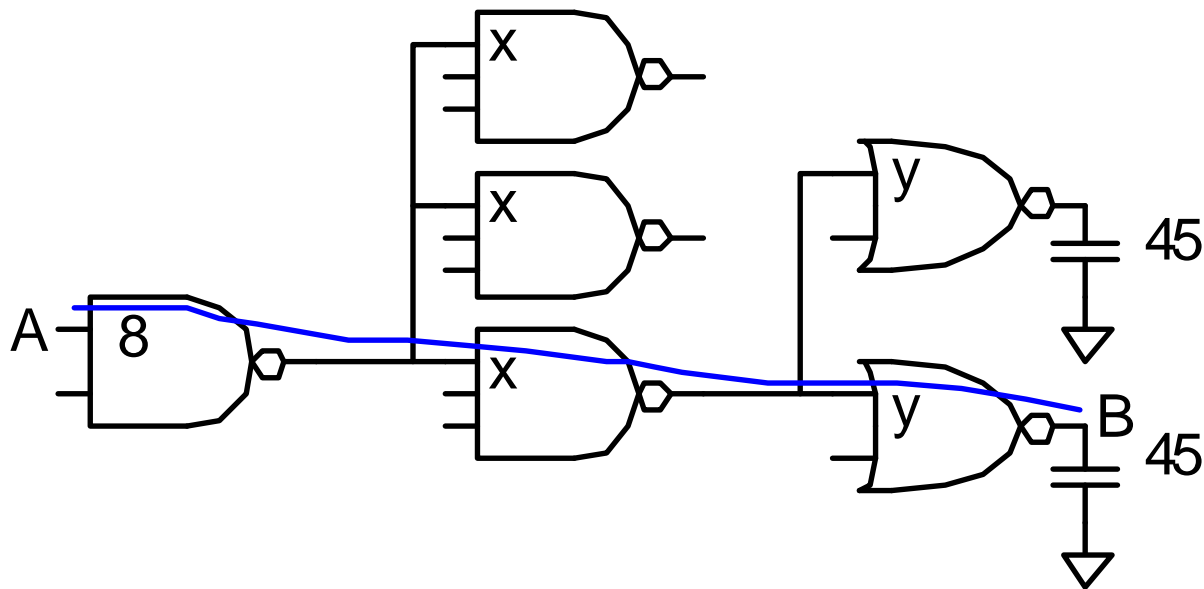
Logical Effort  $G =$   
Electrical Effort  $H =$   
Branching Effort  $B =$   
Path Effort  $F =$   
Best Stage Effort  $\hat{f} =$   
Parasitic Delay  $P =$   
Delay  $D =$

# Example: 3-stage path

- Work backward for sizes

$y =$

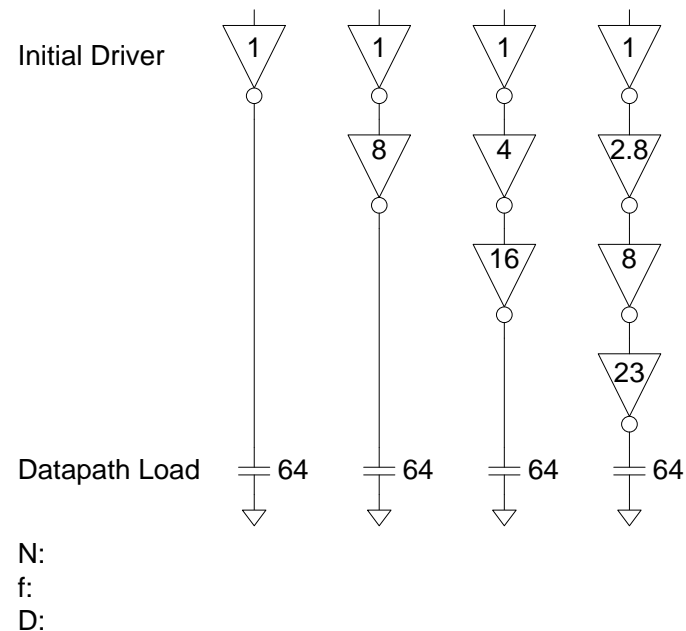
$x =$



# Best Number of Stages

- ❑ How many stages should a path use?
  - Minimizing number of stages is not always fastest
- ❑ Example: drive 64-bit datapath with unit inverter

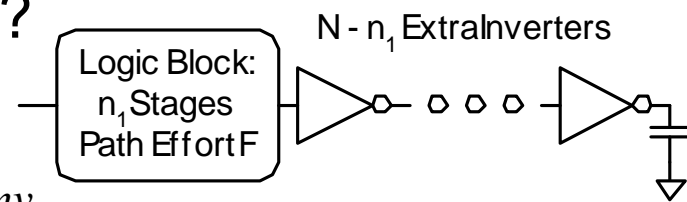
D =



# Derivation

- ❑ Consider adding inverters to end of path
  - How many give least delay?

$$D = NF^{\frac{1}{N}} + \sum_{i=1}^{n_1} p_i + (N - n_1) p_{inv}$$



$$\frac{\partial D}{\partial N} = -F^{\frac{1}{N}} \ln F^{\frac{1}{N}} + F^{\frac{1}{N}} + p_{inv} = 0$$

- ❑ Define best stage effort  $\rho = F^{\frac{1}{N}}$

$$p_{inv} + \rho(1 - \ln \rho) = 0$$

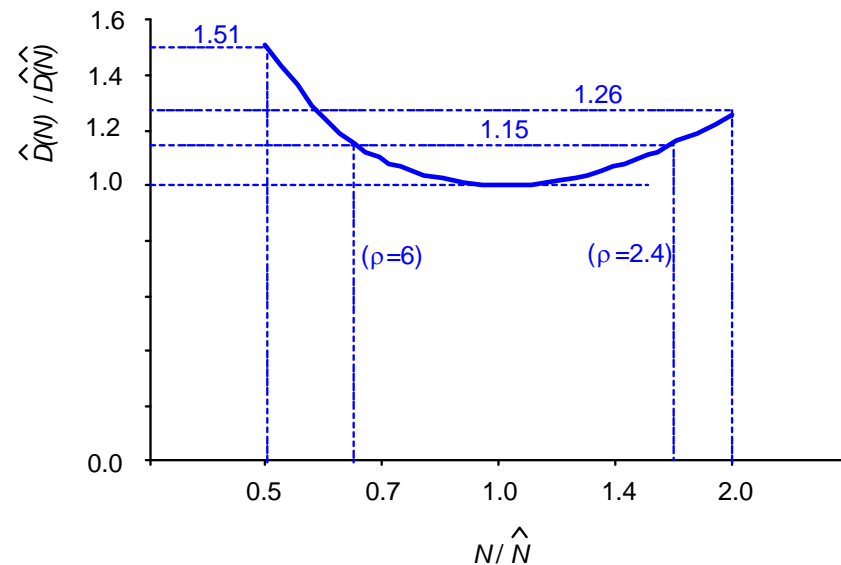
# Best Stage Effort

- ❑  $p_{inv} + \rho(1 - \ln \rho) = 0$  has no closed-form solution
- ❑ Neglecting parasitics ( $p_{inv} = 0$ ), we find  $\rho = 2.718$  (e)
- ❑ For  $p_{inv} = 1$ , solve numerically for  $\rho = 3.59$



# Sensitivity Analysis

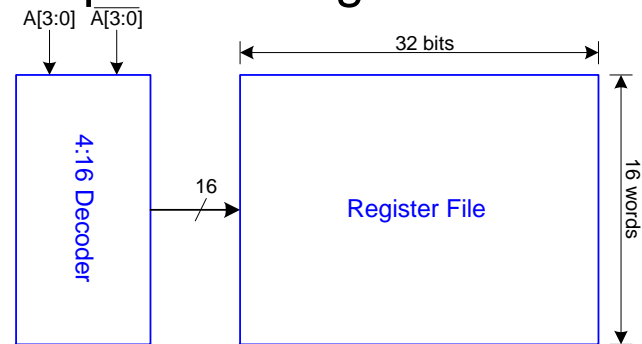
- How sensitive is delay to using exactly the best number of stages?



- $2.4 < \rho < 6$  gives delay within 15% of optimal
  - We can be sloppy!
  - I like  $\rho = 4$

# Example, Revisited

- ❑ Ben Bitdiddle is the memory designer for the Motorola 68W86, an embedded automotive processor. Help Ben design the decoder for a register file.



- ❑ Decoder specifications:
  - 16 word register file
  - Each word is 32 bits wide
  - Each bit presents load of 3 unit-sized transistors
  - True and complementary address inputs A[3:0]
  - Each input may drive 10 unit-sized transistors
- ❑ Ben needs to decide:
  - How many stages to use?
  - How large should each gate be?
  - How fast can decoder operate?

# Number of Stages

- ❑ Decoder effort is mainly electrical and branching

Electrical Effort:  $H =$

Branching Effort:  $B =$

- ❑ If we neglect logical effort (assume  $G = 1$ )

Path Effort:  $F =$

Number of Stages:  $N =$

- ❑ Try a -stage design

# Gate Sizes & Delay

Logical Effort:  $G =$

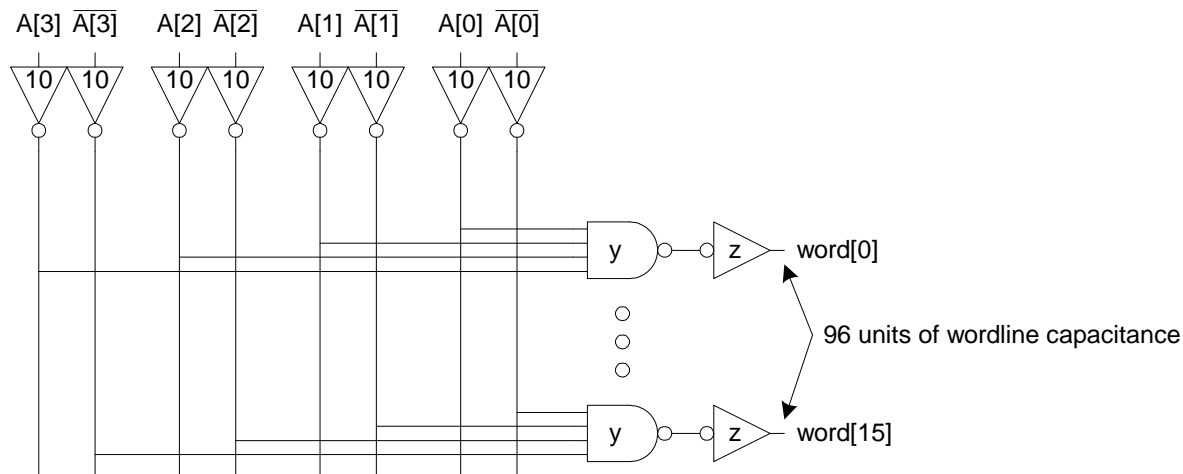
Path Effort:  $F =$

Stage Effort:  $\hat{f} =$

Path Delay:  $D =$

Gate sizes:  $z =$

$y =$



# Comparison

- ❑ Compare many alternatives with a spreadsheet
- ❑  $D = N(76.8 G)^{1/N} + P$

Design	N	G	P	D
NOR4	1	3	4	234
NAND4-INV	2	2	5	29.8
NAND2-NOR2	2	20/9	4	30.1
INV-NAND4-INV	3	2	6	22.1
NAND4-INV-INV-INV	4	2	7	21.1
NAND2-NOR2-INV-INV	4	20/9	6	20.5
NAND2-INV-NAND2-INV	4	16/9	6	19.7
INV-NAND2-INV-NAND2-INV	5	16/9	7	20.4
NAND2-INV-NAND2-INV-INV-INV	6	16/9	8	21.6

# Review of Definitions

Term	Stage	Path
number of stages	1	$N$
logical effort	$g$	$G = \prod g_i$
electrical effort	$h = \frac{C_{out}}{C_{in}}$	$H = \frac{C_{out-path}}{C_{in-path}}$
branching effort	$b = \frac{C_{on-path} + C_{off-path}}{C_{on-path}}$	$B = \prod b_i$
effort	$f = gh$	$F = GBH$
effort delay	$f$	$D_F = \sum f_i$
parasitic delay	$p$	$P = \sum p_i$
delay	$d = f + p$	$D = \sum d_i = D_F + P$

# Method of Logical Effort

- 1) Compute path effort
- 2) Estimate best number of stages
- 3) Sketch path with N stages
- 4) Estimate least delay
- 5) Determine best stage effort
- 6) Find gate sizes

$$F = GBH$$

$$N = \log_4 F$$

$$D = NF^{\frac{1}{N}} + P$$

$$\hat{f} = F^{\frac{1}{N}}$$

$$C_{in_i} = \frac{g_i C_{out_i}}{\hat{f}}$$

# Limits of Logical Effort

- ❑ Chicken and egg problem
  - Need path to compute  $G$
  - But don't know number of stages without  $G$
- ❑ Simplistic delay model
  - Neglects input rise time effects
- ❑ Interconnect
  - Iteration required in designs with wire
- ❑ Maximum speed only
  - Not minimum area/power for constrained delay



# Summary

- ❑ Logical effort is useful for thinking of delay in circuits
  - Numeric logical effort characterizes gates
  - NANDs are faster than NORs in CMOS
  - Paths are fastest when effort delays are  $\sim 4$
  - Path delay is weakly sensitive to stages, sizes
  - But using fewer stages doesn't mean faster paths
  - Delay of path is about  $\log_4 F$  FO4 inverter delays
  - Inverters and NAND2 best for driving large caps
- ❑ Provides language for discussing fast circuits
  - But requires practice to master